Bronze Tech

Single-Channel Isolated Gate Driver

## BTD5350x Single-Channel Isolated Gate Driver

#### 1.Features

- Isolation voltage up to 5000Vrms(SOW-8);
  3000Vrms(SOP-8)
- Peak output current up to 10A
- Propagation delay 60ns
- Maximum switching frequency 1MHz
- Primary-side supply 3~18V
- Secondary-side supply up to 33V
- Primary-side and secondary-side power supply undervoltage lockout (UVLO)
- Compatible with 3.3V, 5V and 15V input
- Feature Options: Miller Clamp Options (BTD5350M)
   Split Outputs (BTD5350S)
   UVLO with respect to IGBT emitter (BTD5350E)
- 8-pin Package: SOP-8 with 4mm creepage SOW-8 with 8.5mm creepage
- Operating Temperature -40~125°C

## 2.Applications

- Motor drives
- EV chargers
- Telecommunication power supplies
- EV power supplies
- UPS
- String solar inverters

# BTD5350x is a family of single-channel, isolated gate driver with peak output current up to 10A, available in

**3.Description** 

SOP-8 (narrow-body) or SOW-8 (wide-body) packages and supports isolation voltages up to 3000Vrms and 5000Vrms respectively. They can be used to drive IGBTs and Si/SiC MOSFETs. BTD5350x family offers 3 Feature Options:

BTD5350M provides Miller clamp function to prevent false turn-on caused by Miller current.

BTD5350S provides a split output with rise and fall time individually configurable.

BTD5350E provides UVLO in positive power supply of secondary-side to ensure that power devices get sufficient gate turn-on voltage.



## 4.Functional Block Diagram



BTD5350M



BTD5350S

#### 5 - VCC2 VCC1 7- GND2 UVLO IN+ -2-> UVLO Level Shift and Ctrl Logic 늡 and input HH OUT Logio SOLATION IN- -3H GND1 - 4 8- VEE2

BTD5350E

## **Bronze Tech** –

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## **5.Product Information**

Part No.	Pin Configuration	Isolation Voltage	UVLO Threshold	Operating Temperature	Package	Package Material	Quantity	Marking		
BTD5350MBPR		3000Vrms	8V		60 D 0		2500pcs	BTD5350MB		
BTD5350MCPR	Miller classes	300071115	11V		SOP-8		/Reel	BTD5350MC		
BTD5350MBWR	Miller-clamp	E000V/rmc	8V		50W 9		1000pcs	BTD5350MB		
BTD5350MCWR		5000Vrms	11V		SOW-8		/Reel	BTD5350MC		
BTD5350SBPR		2000)////////	8V		500.0		2500pcs	BTD5350SB		
BTD5350SCPR	Calitautaut	3000Vrms	11V	-40~125°C	SOP-8	Tape & Reel	/Reel	BTD5350SC		
BTD5350SBWR	Split output	E000)//////	8V	-40~125 C	SOW-8		Reel	Reel	Reel	1000pcs
BTD5350SCWR		5000Vrms	11V				/Reel	BTD5350SC		
BTD5350EBPR		2000)///////	8V		500.0		2500pcs	BTD5350EB		
BTD5350ECPR	UVLO with refer-	3000Vrms	11V		SOP-8		/Reel	BTD5350EC		
BTD5350EBWR	ence to GND2	E000)/##86	8V		50W 9		1000pcs	BTD5350EB		
BTD5350ECWR		5000Vrms	11V		SOW-8		/Reel	BTD5350EC		

## 6.Pin Configuration and Functions

#### 6.1 BTD5350M

NO.	NAME	TYPE <sup>(1)</sup>	DESCRIPTION	PACKAGE
1	VCC1	Р	Input supply voltage	
2	IN+	Ι	Non-inverting gate signal input pin	
3	IN-	Ι	Inverting gate signal input pin	
4	GND1	G	Input ground	IN+ 2 7 CLAMP
5	VCC2	Р	Positive output supply rail	
6	OUT	0	Gate-drive output	GND1 4 5 VCC2
7	CLAMP	I	Miller-clamp input	
8	VEE2	Р	Negative output supply rail	
(1) P=Powe	er, G=Ground	d, I=Input, O	=Output	

#### 6.2 BTD5350S

NO.	NAME	TYPE <sup>(1)</sup>	DESCRIPTION	PACKAGE
1	VCC1	Р	Input supply voltage	
2	IN+	I	Non-inverting gate signal input pin	
3	IN-	I	Inverting gate signal input pin	
4	GND1	G	Input ground	
5	VCC2	Р	Positive output supply rail	IN- 3 6 OUTH
6	OUTH	0	Gate-Drive pullup output pin	GND1 4 5 VCC2
7	OUTL	0	Gate-Drive pulldown output pin	
8	VEE2	Р	Negative output supply rail	
(1) P=Powe	er, G=Ground	d, I=Input, O	=Output	

#### 6.3 BTD5350E

NO.	NAME	TYPE <sup>(1)</sup>	DESCRIPTION	PACKAGE				
1	VCC1	Р	Input supply voltage					
2	IN+	I	Non-inverting gate signal input pin					
3	IN-	I	Inverting gate signal input pin					
4	GND1	G	Input ground	IN+ 2 7 GND2				
5	VCC2	Р	Positive output supply rail					
6	OUT	0	Gate-Drive pullup output pin	GND1 4 5 VCC2				
7	GND2	G	Gate-Drive common pin					
8	VEE2	Р	Negative output supply rail					
(1) P=Powe	(1) P=Power, G=Ground, I=Input, O=Output							

## 7. Specification Parameters

#### 7.1 Absolute Limits

PARAMETER	TEST CONDITIONS	MIN	МАХ	UNIT
Input bias pin supply voltage	VCC1-GND1	-0.3	18	
Driver bias supply	VCC2-VEE2	-0.3	35	]
V <sub>EE2</sub> bipolar supply voltage for E version	VEE2-GND2	-17.5	0.3	V
Secondary side signal voltage	OUT, OUTH, OUTL, CLAMP	VEE2-0.3	VCC2+0.3	]
Primary side signal voltage	IN+, IN-	GND1-5	VCC1+0.3	1
Junction temperature, TJ	-	-40	150	
Storage temperature, T <sub>s</sub>	-	-65	150	°C
Pin soldering temperature, T∟	Duration < 10s	-	300	]
	Human body model (HBM)	±3	000	
ESD voltage	Charged device model (CDM)	±1	35 0.3 VCC2+0.3 VCC1+0.3 150 150	- V

Note: These are stress ratings only, which do not imply functional operation of the device at these or any other conditions beyond those indicated under Recommended Operating Conditions. Exposure to absolute-maximum-rated conditions for extended periods may affect device reliability, and cause permanent damage to the device under severe conditions.

### 7.2 Thermal Information

SYMBOL	PARAMETER	VAI	UNIT	
	PARAMEIER	SOP-8	SOW-8	
R <sub>0JA</sub>	Junction-to-ambient thermal resistance	87.7	91.4	
R <sub>0JC(top)</sub>	Junction-to-case (top) thermal resistance	19.4	35.1	
Rejb	Junction-to-board thermal resistance	43.0	73.8	°C /W
τιψ	Junction-to-top characterization parameter	57.3	51.0	
ψ <sub>ЈВ</sub>	Junction-to-board characterization parameter	65.7	47.7	

#### 7.3 Recommended Operating Conditions

SYMBOL	PARAMETER	PARAMETER			UNIT
V <sub>CC1</sub>	out supply voltage (VCC1–GND1)		3	15	
	Total supply valtage output side (VCC2, VEC2)	D5350xCx	13.2	33	
V <sub>CC2</sub>	Total supply voltage output side (VCC2–VEE2) BTD5350xBx	D5350xBx	9.5	33	V
V <sub>EE2</sub>	Bipolar supply voltage for E version (VEE2–GND2)	-16	0		
TA	Ambient temperature		-40	125	°C

#### 7.4 Power Ratings

PARAMETER	TEST CONDITIONS	MIN	МАХ	UNIT				
SOP-8								
Maximum power dissipation on input and output	V <sub>cc1</sub> =15V, V <sub>cc2</sub> =15V, f=2.1MHz,	-	1.14					
Maximum input power dissipation	50% duty cycle, square wave,	-	0.05	W				
Maximum output power dissipation	2.2nF capacitor	-	1.09	]				
	SOW-8							
Maximum power dissipation on input and output	V <sub>cc1</sub> =15V, V <sub>cc2</sub> =15V, f=1.9MHz,	-	1.04					
Maximum input power dissipation	50% duty cycle, square wave,	-	0.05	W				
Maximum output power dissipation	2.2nF load capacitor	-	0.99	]				

#### **7.5 Electrical Characteristics**

 $T_{\text{A}} {=} {-} 40 {\sim} 125^{\circ} C$  ,  $V_{\text{CC1}} {=} 3.3$  or 5V,  $V_{\text{CC2}} {=} 15 \text{V},$   $C_{\text{L}}{}^{(1)} {=} 100 \text{pF}.$ 

Output pin: current towards outside of the chip is positive direction; Input pin: current towards inside of the chip is positive direction.

SYMBOL	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
	SUPPL	Y CURRENTS				
I <sub>VCC1</sub>	Input supply quiescent current	-	-	1.4	2.4	
Ivcc2	Output supply quiescent current	-	-	1.3	1.8	mA
	SUPPLY VOLTAGE UN	DERVOLTAGE THRESHO	LDS			
Voni	VCC1-GND1 Positive-going UVLO threshold voltage	-	-	2.6	2.8	
Voff1	VCC1-GND1 Negative-going UVLO threshold voltage	-	2.4	2.5	-	V
V <sub>UV, HYS1</sub>	VCC1-GND1 UVLO threshold hysteresis	-	-	0.1	2.4 1.8 2.8  - 13 - 13 - 9.4 - 0.7×Vcc1 cc1 0.7×Vcc1 cc1 -	]
	UVLO THRESI	HOLDS BTD5350xCx				
V <sub>ON2</sub>	VCC2-VEE2 Positive-going UVLO threshold voltage	-	-	12	13	
V <sub>OFF2</sub>	VCC2-VEE2 Negative-going UVLO threshold voltage	-	10.3	11	-	] v
V <sub>UV, HYS2</sub>	VCC2-VEE2 UVLO threshold voltage hysteresis	-	-	1	-	1
	UVLO THRESH	HOLDS BTD5350xBx				
V <sub>ON2</sub>	VCC2-VEE2 Positive-going UVLO threshold voltage	-	-	8.7	9.4	
V <sub>OFF2</sub>	VCC2-VEE2 Negative-going UVLO threshold voltage	-	7.3	8	-	V
V <sub>UV, HYS2</sub>	VCC2-VEE2 UVLO threshold voltage hysteresis	-	-	0.7	-	1
	L	DGICI/O		1		1
VIH	Positive-going input threshold voltage (IN+, IN-)	-	-	0.55×V <sub>CC1</sub>	0.7×V <sub>CC1</sub>	
VIL	Negative-going input threshold voltage (IN+, IN-)	-	0.35×V <sub>CC1</sub>	0.45×V <sub>CC1</sub>	-	l v
VIN, HYS	Input Hysteresis voltage	-	-	0.1×V <sub>CC1</sub>	-	1
Ьн	High-level input leakage at IN+	IN+=V <sub>CC1</sub> , V <sub>CC1</sub> =5V	-	55	240	
		IN-=GND1, V <sub>CC1</sub> =5V	-240	-54	-	μA
$ _{ L}$	Low-level input leakage at IN-	IN-=GND1-5V, Vcc1=5V	-310	-108	-	
	GATE D	RIVER STAGE				1
I <sub>он</sub>	Peak source current	IN+=HIGH, IN-=LOW, C∟=470nF	-	10	-	
Iol	Peak sink current	IN+=LOW, IN-=HIGH, C∟=470nF	-	10	-	A
Vон	High level output voltage VCC2-OUT or VCC2-OUTH	I₀u⊤=+20mA IN+=HIGH, IN-=LOW	-	60	-	mV
Vol	Low level output voltage OUT-VEE2 or OUTL-VEE2	I <sub>out</sub> =-20mA IN+=LOW, IN-=HIGH	5	7	-	mV
	MILLER CL	AMP BTD5350Mx				
$V_{clamp}$	Low-level clamp voltage	I <sub>clamp</sub> =20mA	-	7	10	mV
I <sub>clamp</sub>	Low-level clamp current	V <sub>CLAMP</sub> =VEEx+15V	5	10	-	A
V <sub>clamp-TH</sub>	Clamping threshold voltage	-	-	2.2	-	V
	SHORT-CIF	RCUIT CLAMPING				
	Clamping voltage (OUT-VCC2 or OUTH-VCC2)	IN+=HIGH, IN–=LOW, t <sub>clamp</sub> =10µs, Iouth or Iout=-500mA	-	1	1.3	
Vclp-out	Clamping voltage (VEE2-OUT or VEE2-OUTL or VEE2-CLAMP)	IN+=LOW, IN-=HIGH, t <sub>clamp</sub> =10μs, I <sub>clamp</sub> or I <sub>outl</sub> =500mA	-	1.5	-	V
		IN+=LOW, IN-=HIGH, I <sub>CLAMP</sub> or I <sub>OUTL</sub> =20mA	-	0.9	1	
	ACTIVE PULL	-DOWN FUNCTION				
Voutsd	Active pulldown voltage on OUTL, CLAMP, OUT	I <sub>ouτ</sub> =-1A (sinking into OUT, OUTL or CLAMP pin), VCC2=left open	-	2.3	2.6	v



(Continued)

		5	SWITCHING PARAMETERS				
t <sub>PLH</sub>	Propagation delay, high		CL=100pF	-	60	75	
t <sub>PHL</sub>	Propagation delay, low		CL=100pF	-	60	75	1
tr	Output-signal rise time		CL=1nF	-	10	26	]
tr	Output-signal fall time		CL=1nF	-	10	26	ns
t <sub>PWD</sub>	Pulse width distortion[t <sub>PHL</sub> -t <sub>PLH</sub> ]		CL=100pF	-	1	20	]
t <sub>sk</sub> 2)	Part-to-part skew		CL=100pF	-	1	25	]
t <sub>UVL01-rec</sub>	Undervoltage lockout	Input side	VCC1	-	50	-	
t <sub>UVLO2-rec</sub>	recovery delay time	Output side	VCC2	-	50	-	μs
CMTI	Common-mode transient immunity		INx tied to GND or VCC1, V <sub>CM</sub> =1500V	150	-	-	kV/μs

Note: 1.  $C_L$ : Load capacitance from output pin OUTx to VEE2 or GND2.

2. t<sub>sk</sub> is the magnitude of the difference in propagation delay times between the output of different devices switching in the same direction while operating at identical supply voltages, temperature, input signals and loads guaranteed by characterization.

### 7.6 Insulation Specifications

#### 7.6.1 SOP-8 Safety Parameters

SYMBOL	PARAMETER	TEST CONDITIONS	MIN	TYP	МАХ	UNIT	
CLR	External Clearance	-	4	-	-		
CPG	External Creepage	-	4	-	-	mm	
DTI	Distance through the insulation	-	17	-	-	μm	
CTI	Comparative tracking index	DIN EN 60112	600	-	-	V	
	Overvoltage category per IEC 60664-1	Rated mains voltage ≤ 150Vrms	I-IV	-	-		
	Overvoltage category per IEC 60664-1	Rated mains voltage ≤ 300Vrms	-	-	-	_	
		DIN VDE V0884-11					
VIORM	Maximum repetitive peak isolation voltage	AC voltage (bipolar)	990	-	-	Vpk	
VIOWM	Maximum isolation working voltage	AC voltage (sine wave); time dependent dielectric breakdown (TDDB) test	700	-	-	Vrms	
VIOTM	Maximum transient isolation voltage	$V_{\text{TEST}}=V_{\text{IOTM}}=V_{\text{ISO}}$ , t=60s (qualification) ; $V_{\text{TEST}}=V_{\text{IOTM}}=1.2 \times V_{\text{ISO}}$ , t=1s (100% production)	4242	-	-	- V <sub>PK</sub>	
VIOSM	Maximum surge isolation voltage	Test method per IEC 62368-1, 1.2/50- $\mu$ s waveform, V <sub>TEST</sub> =1.3×V <sub>IOSM</sub> (qualification)	4242	-	-		
	Apparent charge	Method a: After I/O safety test subgroup 2/3, $V_{ini}=V_{IOTM}$ , $t_{ini}=60$ s, $V_{pd(m)}=1.2 \times V_{IOTM}$ , $t_m=10$ s	-	-	5		
q <sub>pd</sub>		$\begin{array}{l} \mbox{Method} a: \mbox{After environmental tests subgroup 1,} \\ \mbox{V}_{ini} = \mbox{V}_{IOTM}, \mbox{t}_{ini} = 60 s, \mbox{V}_{pd(m)} = 1.2 \times \mbox{V}_{IOTM}, \mbox{t}_{m} = 10 s \end{array}$	-	-	5	рС	
		Method b1: At routine test (100% production) and preconditioning (type test), $V_{ini}$ =1.2× $V_{IOTM, t_{ini}$ =1s, $V_{pd(m)}$ =1.5× $V_{IORM, t_m}$ =1s	-	-	5		
CIO	Barrier capacitance, input to output	V10=0.4×sin(2πft), f=1MHz	-	1.2	-	рF	
		Test voltage of 500V, T <sub>A</sub> =25°C	10 <sup>12</sup>	-	-		
Rio	Isolation resistance, input to output	Test voltage of 500V, 100°C $\leqslant$ $T_{A}$ $\leqslant$ 125°C	1011	-	-	Ω	
		Test voltage of 500V, T <sub>s</sub> =150°C	10 <sup>9</sup>	-	-	1	
	Pollution degree	-	-	2	-	-	
		UL1577					
V <sub>ISO</sub>	Withstand isolation voltage	$V_{\text{TEST}}=V_{\text{ISO}}$ , t=60s (qualification) ; $V_{\text{TEST}}=1.2 \times V_{\text{ISO}}$ , t=1s (100% production)	3000	-	-	Vrms	



#### 7.6.2 SOW-8 Safety Parameters

SYMBOL	PARAMETER	TEST CONDITIONS	MIN	TYP	MAX	UNIT
CLR	External Clearance	-	8.5	-	-	
CPG	External Creepage	-	8.5	-	-	mm
DTI	Distance through the insulation	-	17	-	-	μm
CTI	Comparative tracking index	DIN EN 60112	600	-	-	V
		Rated mains voltage ≤ 600Vrms	-	-	-	
	Overvoltage category per IEC 60664-1	Rated mains voltage ≤ 1000Vrms	-	-	-	-
		DIN V VDE 0884–11				
VIORM	Maximum repetitive peak isolation voltage	AC voltage (bipolar)	2121	-	-	Vpk
VIOWM	Maximum isolation working voltage	AC voltage (sine wave); time dependent dielectric breakdown (TDDB) test		-	-	Vrms
VIOTM	Maximum transient isolation voltage	$ \begin{array}{l} V_{\text{TEST}} = V_{\text{IOTM}} = V_{\text{ISO}}, t = 60 \text{s} \left( \text{qualification} \right); \\ V_{\text{TEST}} = V_{\text{IOTM}} = 1.2 \times V_{\text{ISO}}, t = 1 \text{s} \left( 100\% \text{ production} \right) \end{array} $	7000		-	- Урк
VIOSM	Maximum surge isolation voltage	Test method per IEC 62368-1, 1.2/50- $\mu$ s waveform, V <sub>TEST</sub> =1.6 × V <sub>IOSM</sub> (qualification)	8000	-	-	VPK
		Method a: After I/O safety test subgroup 2/3, $V_{ini}=V_{IOTM}$ , $t_{ini}=60s$ , $V_{pd(m)}=1.2 \times V_{IOTM}$ , $t_m=10s$	-	-	5	
q <sub>pd</sub>	Apparent charge	Method a: After environmental tests subgroup 1, $V_{ini}=V_{IOTM}$ , $t_{ini}=60$ s, $V_{pd(m)}=1.6 \times V_{IORM}$ , $t_m=10$ s	-	-	5	pC
		Method b1: At routine test (100% production) and preconditioning (type test), $V_{ini}$ =1.2× $V_{IOTM,}$ t <sub>ini</sub> =1s, $V_{pd(m)}$ =1.875× $V_{IORM}$ , t <sub>m</sub> =1s	-	-	5	
CIO	Barrier capacitance, input to output	V <sub>I0</sub> =0.4×sin(2πft), f=1MHz	-	1.2	-	pF
		Test voltage of 500V, T <sub>A</sub> =25°C	1012	-	-	
Rio	Isolation resistance, input to output	Test voltage of 500V, 100°C $\leqslant$ $T_{A}$ $\leqslant$ 125°C	1011	-	-	Ω
		Test voltage of 500V, T <sub>s</sub> =150°C	10 <sup>9</sup>	-	-	1
	Pollution degree	-	-	2	-	-
		UL1577				
Viso	Withstand isolation voltage	$V_{TEST}=V_{ISO}$ , t=60s (qualification) ; $V_{TEST}=1.2 \times V_{ISO}$ , t=1s (100% production)	5000	-	-	Vrms

## 7.6.3 SOW-8 Packing Safety-Related Certifications

UL	VDE	CQC
Recognized under UL 1577 Component Recognition Program	Plan to certify according to DIN V VDE V0884-11:2017-01 and DIN EN 61010-1	Plan to certify according to GB 4943.1-2011
Single protection, 5000 V <sub>RMS</sub>	Reinforced Insulation Maximum Transient isolation Overvoltage, 7000 V <sub>PK</sub> ; Maximum Repetitive Peak Isolation Voltage, 2121 V <sub>PK</sub> ; Maximum Surge Isolation Voltage, 8000 V <sub>PK</sub>	Reinforced Insulation, Altitude ≤ 5000 m, Tropical Climate
File Number: E537161	Certification planned	Certification planned



## 8.Parameter Testing

#### **8.1 Propagation Delay**

The method for measuring the rise time  $(t_r)$  and fall time  $(t_f)$ : the propagation delay OUTH and OUTL for non-inverting input (see Figure 1); the propagation delay for inverting input (see Figure 2).





## **9.Function Description**

#### 9.1 Isolation Design Description

The isolation inside the BTD5350x series is implemented with high voltage SiO<sub>2</sub> based capacitors. There is a capacitor on each of the primary-side and secondary-side to enhance insulation. The signal across the isolation has an on-off keying (OOK) modulation scheme to transmit the digital data. The primary-side transmits a high-frequency carrier to represent one digital state, and sends no signal to represent the other digital state. On the secondary side, the receiver demodulates the signal and produces output for control. One can also add special anti-jamming circuit on the primary /secondary sides to enhance the CMTI performance and minimize the radiate emissions (see Figure 6, 7).



Figure 6. Conceptual Block Diagram of a Capacitive Data Channel



Figure 7. OOK Based Modulation Scheme

#### 9.2 Input Stage Characteristics

With input pins and secondary side completely isolated, BTD5350x is designed to be CMOS-compatible. It supports 3.3V, 5V and 15V level input, making the chip easy to accept control of multiple logic levels. It also has a Internal integrated filter circuit with 0.1\* VCC1 hysteresis for improving noise immunity of the input stage. The IN+ input port has a 128k $\Omega$  pull-down resistor to force it grounding, the IN- input port has a 128k $\Omega$  pull-up resistor to make it go high to the power supply, which can ensure that the output ports are in OFF state when input ports are left open. However, in order to configure the initial power-on state of the driver IC, Bronze Technologies still recommends adding externally an appropriate pull-up or pull-down resistor to the input.

#### 9.3 Output Booster Characteristics

The BTD5350x has a rail-to-rail push stage output. The pull-up structure of the output stage consists of a P-channel MOSFET and an N-channel MOSFET connected in parallel. At turn-on, N-channel MOSFET provides high current driving capability. P-channel MOSFET provides a small steady-state conduction voltage drop during steady conduction. The pull-down structure is implemented using an N-channel MOSFET. A  $1M\Omega$  resistor is connected in parallel between the drain and gate of the MOSFET to effectively clamp the gate voltage of the power device in the event of a loss of power on the chip to prevent the occurrence of partial turn-on. However, in order to ensure reliable shutdown of the power device, Bronze recommends that appropriate pull-down resistor be added to the gate.



In order to prevent the output port voltage from overvoltage/overcurrent damage when the positive secondary-side power supply voltage VCC2 becomes lower than the negative secondary-side power supply voltage VEE2, due to gate oscillation on output terminals, the 5350M version is designed with a Miller clamp on the output port to effectively suppress oscillations. However, to ensure the normal operation of the drive circuit, Bronze recommends to use Schottky diodes at the gate to clamp the positive and negative power supplies respectively.



Figure 8. Output Stage

#### **9.4 Protection Functions**

#### 9.4.1 Undervoltage Lockout

The BTD5350x has undervoltage lockout (UVLO) function on the primary-side power supply and the secondary-side total supply voltage to prevent the gate drive voltage from being insufficient. When the supply voltage drops below UVLO threshold, the ASIC turns off the output to protect the power devices. When the supply voltage reaches the positive-going negative-going threshold, the ASIC resumes the output. To prevent repeated action near the UVLO threshold, the ASIC is configured with hysteresis. In order to avoid the uncertainty of the output state after power-on, the ASIC first enters an UVLO state after power-on, keeps the output off until the supply voltage is established and then starts normal operation (see Figure 9).





#### 9.4.2 Miller Clamp

The active Miller clamp function is used to prevent the power devices from false turn-on of the power switched cause by the Miller current. A low impedance path is added between the gate terminal and ground (VEE2) to sink the Miller current. The Miller clamp function clamps the gate voltage of the power device to less than 2V when the output is in the off state.



#### 9.4.3 Short-Circuit Clamping

In short-circuit conditions, the gate voltage of the power devices tend to rise, so the short-circuit clamping function is used to clamp its gate voltage. The short-circuit clamping function helps protect the IGBT or MOSFET gates from overvoltage breakdown or degradation. The short circuit clamping function is implemented by adding a diode between the OUT, OUTL or CLAMP pins and the VCC2 pins inside the driver. The internal diode can carry a current up to 500mA for 10µs or a continuous current of 20mA. Use external Schottky diodes to improve current conduction capability as needed.

#### 9.4.4 Active Pull-Down

The active pull-down function is used to pull the IGBT or MOSFET gate to the low state when no power is connected to the VCC2. This feature prevents false IGBT and MOSFET turn-on on the OUT, OUTL and CLAMP pins by clamping the output to approximately 2V.

#### 9.5 ESD Structure

The figure below shows the ESD-protected diode configuration of the input and output pins.



#### 9.6 Truth Table

BTD5350M and BTD5350E

IN+	IN-	VCC1	VCC2	OUT	
Х	Х	<uvlo< td=""><td>Х</td><td>L</td></uvlo<>	Х	L	
Х	Х	Х	<uvlo< td=""><td>L</td></uvlo<>	L	
Н	L	>UVLO	>UVLO	Н	
Н	Н	>UVLO	>UVLO	L	
L	L	>UVLO	>UVLO	L	
L	Н	>UVLO	>UVLO	L	

BTD5350S

IN+	IN-	VCC1	VCC2	OUTH	OUTL
Х	Х	<uvlo< td=""><td>Х</td><td>Hi-Z</td><td>L</td></uvlo<>	Х	Hi-Z	L
Х	Х	Х	<uvlo< td=""><td>Hi-Z</td><td>L</td></uvlo<>	Hi-Z	L
Н	L	>UVLO	>UVLO	Н	Hi-Z
Н	Н	>UVLO	>UVLO	Hi-Z	L
L	L	>UVLO	>UVLO	Hi-Z	L
L	Н	>UVLO	>UVLO	Hi-Z	L



## **10.Applications**

The following sections introduce the basic typical application of Bronze Technologies driver ICs, which is for reference only. In practical application, users need to verify and test its applicability according to their own design requirements to confirm the system functions.

#### **10.1 Typical Applications**

Bronze Technologies recommends that customers add a RC filter with a small time constant at the input port to filter out high-frequency interference without adding a large delay. It is recommended that the resistance value should be between 0 and  $100\Omega$  and the capacitance should be less than 1000 pF. When setting this parameter, the influence between high frequency interference and delay needs to be taken into account.

To ensure the supply stability, Bronze Technologies recommends to add appropriate blocking capacitor between the power supply and ground. It is recommended to parallel a 1uF+ 0.1uF capacitor between VCC1-GND1, and a 10uF+ 0.22uF capacitor between VCC2-VEE2 (Figure 11,12,13).



Figure 11. Typical Application Circuit for BTD5350M



Figure 12. Typical Application Circuit for BTD5350S



Figure 13. Typical Application Circuit for BTD5350E



#### **10.2 Recommended Design Of Secondary Side Supply**

In order to avoid false turn-on of the gate of the power device due to interference, it is recommended that customers add a negative power supply when designing the driving output. It is recommended to use the following two methods to generate the negative supply: use a regulator to generate stable negative voltage (see Figure 14); or use both positive and negative supplies (see Figure 15).





Figure 14. Voltage Regulator Design

Figure 15. Dual-Supply Design



## **11.**Packaging and Packing Information

#### 11.1 Package Identifier

11.1.1 SOP-8 Package Identifier



Note: 1) Legend unit: mm.



#### 11.1.2 SOW-8 Package Identifier



Note: 1) Legend unit: mm.

#### **Electrostatic Discharge Caution**



This integrated circuit can be damaged by ESD.Bronze Technologies that all integrated circuits be handled with appropriate precautions.Failure to observe proper handling and installation procedures can cause damage.

ESD damage can range from subtle performance degradation to complete device failure.Precision integrated circuits may be more susceptible to damage because very small parametric changes could cause the device not to meet its published specifications.

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#### **11.2 Packing Information**

11.2.1 SOP-8 Packing Information



Note: 1) Legend unit: mm.

**REEL DIMENSIONS** 





ITEM	FOOTPRINT
Reel Diameter	13 inches
Reel Width(W1)	12.4mm



#### 11.2.2 SOW-8 Packing Information



A0(6.45-6.50)

Note: 1) Legend unit: mm.

#### **REEL DIMENSIONS**





ITEM	FOOTPRINT	
Reel Diameter	13 inches	
Reel Width(W1)	16.4mm	



## **12.Version Description**

REVISION	NOTES	DATE
Rev.0.0	Released datasheet	04-Jan-2023
Rev.0.1	Figures and description updated	28-Nov-2023
Rev.0.2	Safety-related certifications added	04-Jun-2024
Rev.0.3	Figures updated	09-Jul-2024

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